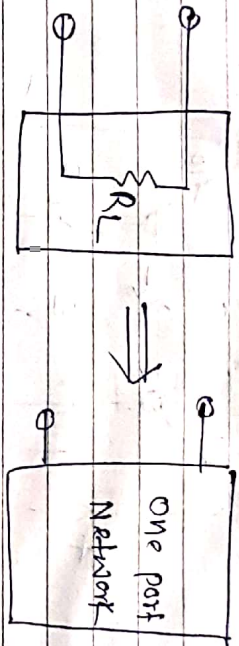


① Two part Network

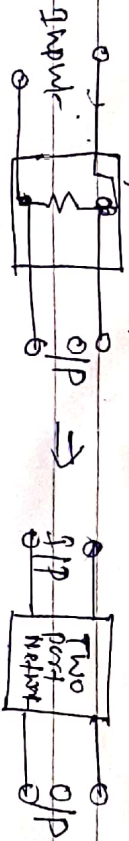
One port



Any active or passive network having only two terminals can be represented by an one port network. It is represented by a rectangular box. It must have two terminals. This rectangular box then represents one port or a two terminal network.

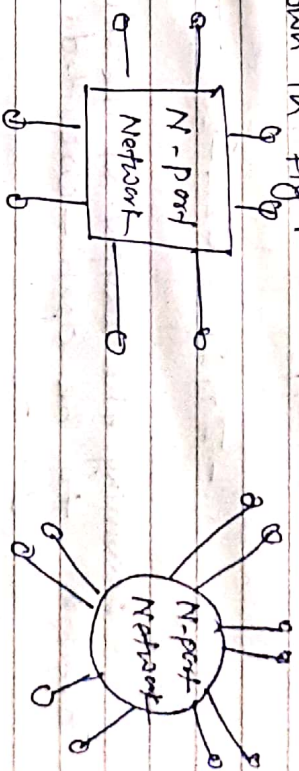
Two Port

If the network box that represents a network consists of two pairs of terminals, where one pair of terminals the other pair being output it is called a two port network. Fig represents the diagrammatic representation of a two port network. In such a network the driving source may be connected at the pair of input terminals while the other pair of terminals serves as output.

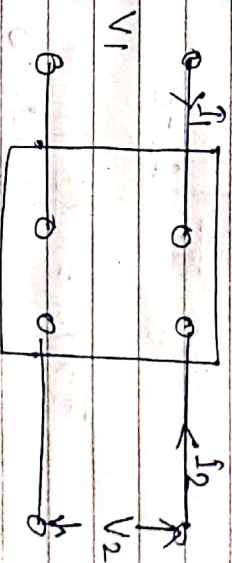


N-port network

If the network representation contains of 'n' number of pairs of terminals, it is called a ~~two~~ n-port network. In a port network, the driving source may be connected in more than one number of pairs of terminals. The block diagram of an ~~network~~ n-port network has been shown in fig.



Z-Parameters (open circuit impedance parameters)



For the two port network in this part V_1 & I_1 are the inputs and V_2 & I_2 are outputs. It can be expressed in terms of

$$V = [Z][I]$$

①

(3)

Where $[Z]$ is the impedance matrix.
Equation (1) can be expressed as

$$\begin{bmatrix} V_1 \\ V_2 \end{bmatrix} = \begin{bmatrix} Z_{11} & Z_{12} \\ Z_{21} & Z_{22} \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \end{bmatrix} \quad (2)$$

By equation (2) gives

$$V_1 = Z_{11}I_1 + Z_{12}I_2 \quad (3)$$

$$V_2 = Z_{21}I_1 + Z_{22}I_2 \quad (4)$$

Assuming the output of the two-port network to be open-circuited.

$$I_2 = 0$$

Then from eqn (3) with

$$I_2 = 0,$$

$$Z_{11} = \frac{V_1}{I_1} \Big|_{I_2=0}$$

$$Z_{21} = \frac{V_2}{I_1} \Big|_{I_2=0}$$

Again, assuming the input port of the same two port network to be open ckt

$$I_1 = 0$$

From eqn (3)

$$Z_{12} = \frac{V_1}{I_2} \Big|_{I_1=0}$$

(4)

and from equation

$$Z_{22} = \frac{V_2}{I_2} \Big|_{I_1=0}$$

Then

$$Z_{11} = \frac{V_1}{I_1} \Big|_{I_2=0} \quad \& \quad Z_{22} = \frac{V_2}{I_2} \Big|_{I_1=0}$$

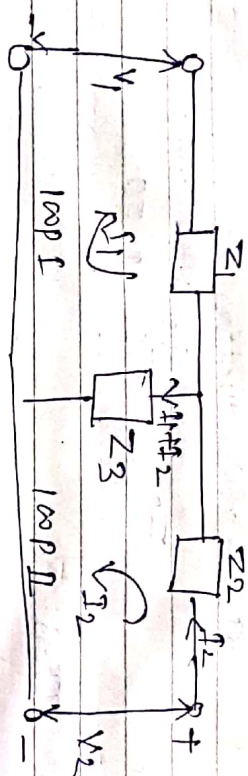
These are called input & o/p driving point impedances respectively.

$$\text{and } Z_{12} = \frac{V_1}{I_2} \Big|_{I_1=0} \quad \& \quad Z_{21} = \frac{V_2}{I_1} \Big|_{I_2=0}$$

are called forward transfer & impedance parameters or open-circuit parameters of the two port network.

(5)

① Find the Z-parameters for the following network



Soln: By KVL in the above fig, the loop eqns can be written as

$$V_1 = I_1 Z_1 + (I_1 + I_2) Z_2$$

$$\Rightarrow V_1 = (Z_1 + Z_2) I_1 + Z_2 I_2 \quad \text{--- (I)}$$

$$\text{and } V_2 = I_2 Z_2 + (I_1 + I_2) Z_3$$

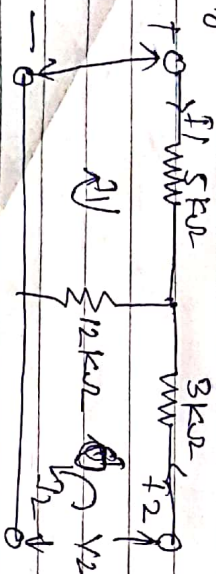
$$\Rightarrow V_2 = Z_3 I_1 + (Z_2 + Z_3) I_2 \quad \text{--- (II)}$$

Comparing eqns (I) & (II) with the standard Z-parameter equation

$$Z_{11} = Z_1 + Z_2, \quad Z_{12} = Z_2$$

$$Z_{21} = Z_3, \quad Z_{22} = Z_2 + Z_3$$

Q:- Find Z-parameter for the network shown in fig.



Using by KVL for loops I & II.

$$V_1 = 5 \times 10^3 I_1 + 12 \times 10^3 (I_1 + I_2)$$

$$\Rightarrow V_1 = 17 \times 10^3 I_1 + 12 \times 10^3 I_2 \quad \text{--- (I)}$$

and

$$V_2 = 3 \times 10^3 I_2 + 12 \times 10^3 (I_1 + I_2)$$

$$\Rightarrow V_2 = 12 \times 10^3 I_1 + 15 \times 10^3 I_2 \quad \text{--- (II)}$$

Comparing these two eqns (I) & (II) with Z-parameters.

$$Z_{11} = 17 \times 10^3 \Omega$$

$$Z_{12} = 12 \times 10^3 \Omega$$

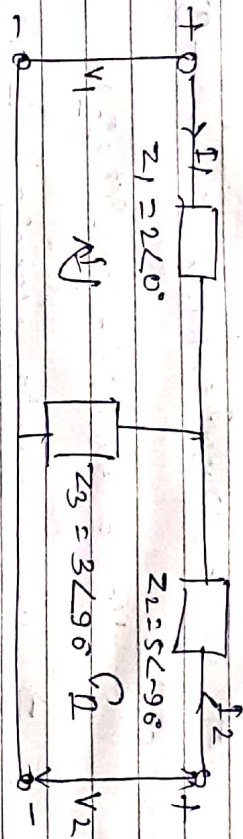
$$Z_{21} = 12 \times 10^3 \Omega$$

$$Z_{22} = 15 \times 10^3 \Omega$$

(6)

(7)

Q:- In a T network of fig. $Z_1 = 2 \angle 0^\circ$, $Z_2 = 5 \angle -90^\circ$, $Z_3 = 3 \angle 90^\circ$, find the Z-parameters.



Soln

By KVL, the loop equations are

$$V_1 = 2 \angle 0^\circ I_1 + 3 \angle 90^\circ (I_1 + I_2)$$

$$\Rightarrow V_1 = [2 \angle 0^\circ + 3 \angle 90^\circ] I_1 + 3 \angle 90^\circ I_2 \quad \text{--- (I)}$$

$$\text{and } V_2 = 5 \angle -90^\circ I_2 + 3 \angle 90^\circ (I_1 + I_2)$$

$$\Rightarrow V_2 = 3 \angle 90^\circ I_1 + [5 \angle -90^\circ + 3 \angle 90^\circ] I_2 \quad \text{--- (II)}$$

From (I) & (II)

$$Z_{11} = 2 \angle 0^\circ + 3 \angle 90^\circ = 2 + j3$$

$$= 3.605 \angle +56.3^\circ \Omega$$

$$Z_{12} = 3 \angle 90^\circ \Omega$$

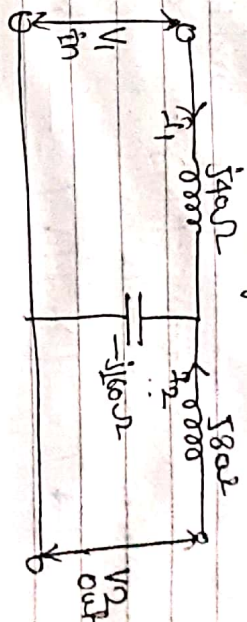
$$Z_{21} = 3 \angle 90^\circ \Omega$$

$$Z_{22} = 5 \angle -90^\circ + 3 \angle 90^\circ$$

$$= -j5 + j3 = -2j \Omega \quad 2 \angle -90^\circ \Omega$$

(8)

Q:- Find the open circuit parameter of the two port network shown in fig



Soln: Assuming open ckt at the output terminals

$$V_1 = I_1 (j40 - j160) \text{ V}$$

$$= -I_1 j120 \Omega$$

$$\text{Then } Z_{11} = \frac{V_1}{I_1} \bigg|_{I_2=0} = -j120 \Omega$$

and

$$\text{since } V_2 = I_1 (-j160)$$

$$\therefore Z_{21} = \frac{V_2}{I_1} \bigg|_{I_2=0} = -j160 \Omega$$

Similarly,

$$V_2 = I_2 j(80 - 160) = -jI_2 80 \text{ V}$$

that gives

$$Z_{22} \big|_{I_1=0} = \frac{V_2}{I_2} = -j80 \Omega$$

$$\text{and } Z_{12} = \frac{V_1}{I_2} \bigg|_{I_1=0} = \frac{-j160 I_2}{I_2} = -j160 \Omega$$

Thus in this network $[Z_{11} = -j120 \Omega, Z_{12} = -j160 \Omega, Z_{21} = -j160 \Omega, Z_{22} = -j80 \Omega]$

(9)

Y-Parameters (short-circuit admittance parameters)

In a two port network, the input current I_1 and I_2 can be expressed in terms of input and output voltages V_1 and V_2 respectively as

$$\begin{bmatrix} I_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} Y_{11} & Y_{12} \\ Y_{21} & Y_{22} \end{bmatrix} \begin{bmatrix} V_1 \\ V_2 \end{bmatrix} \quad \text{--- (I)}$$

Equation (I) gives

$$I_1 = Y_{11}V_1 + Y_{12}V_2 \quad \text{--- (II)}$$

$$I_2 = Y_{21}V_1 + Y_{22}V_2 \quad \text{--- (IV)}$$

Assuming output voltage $V_2 = 0$

From eqn (II) & (IV)

$$Y_{11} = \frac{I_1}{V_1} \Big|_{V_2=0}$$

$$Y_{21} = \frac{I_2}{V_1} \Big|_{V_2=0}$$

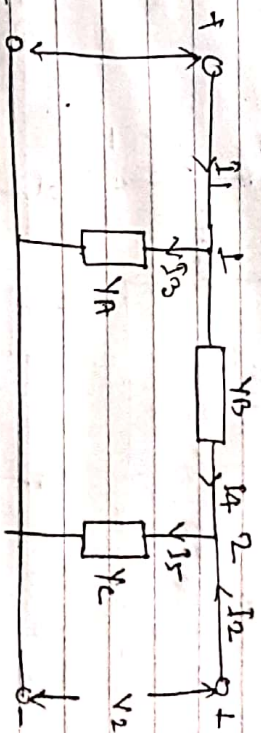
Similarly, assuming the input port to be short ckt, $V_1 = 0$ then eqn (II) & (IV) gives

$$Y_{12} = \frac{I_1}{V_2} \Big|_{V_1=0}$$

$$Y_{22} = \frac{I_2}{V_2} \Big|_{V_1=0}$$

~~Find the Y-parameters of the following~~

Q:- Find the Y-parameters of the following π -ckt and draw



By using KCL at node 1

$$I_1 = I_4 + I_3$$

$$\Rightarrow I_1 = V_1 Y_A + (V_1 - V_2) Y_B$$

$$\Rightarrow I_1 = V_1(Y_A + Y_B) + (-Y_B)V_2 \quad \text{--- (I)}$$

Now using KCL at node 2

$$I_2 = I_5 - I_4$$

$$\Rightarrow I_2 = V_2 Y_C - (V_1 - V_2) Y_B$$

$$\Rightarrow I_2 = (-Y_B)V_1 + (Y_C + Y_B)V_2 \quad \text{--- (II)}$$

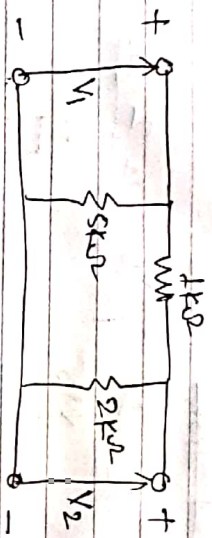
comparing eqn (I) & (II)

$$Y_{11} = Y_A + Y_B, \quad Y_{12} = -Y_B$$

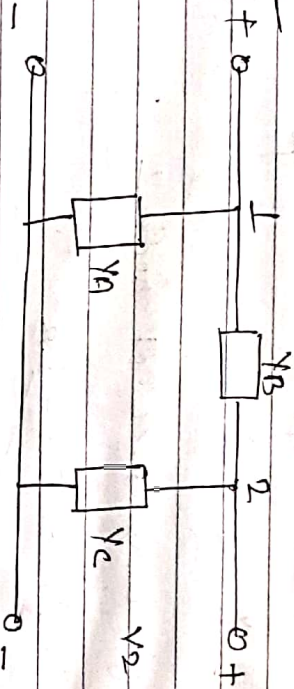
$$Y_{21} = -Y_B, \quad Y_{22} = Y_C + Y_B$$

Q:- An π -attenuator has been shown in fig.

Find the Y -parameters.



Soln.



$$Y_A = \frac{1}{5k\Omega} = 0.2 \times 10^{-3} \Omega^{-1}$$

$$Y_B = \frac{1}{1k\Omega} = 10^{-3} \Omega^{-1}$$

$$Y_C = \frac{1}{2k\Omega} = 0.5 \times 10^{-3} \Omega^{-1}$$

The Y -parameters are

$$Y_{11} = Y_A + Y_B = 1.2 \times 10^{-3} \Omega^{-1}$$

$$Y_{12} = -10^{-3} \Omega^{-1}$$

$$Y_{21} = -10^{-3} \Omega^{-1}$$

$$Y_{22} = Y_B + Y_C = 1.5 \times 10^{-3} \Omega^{-1}$$

Hybrid parameter

As both short-circuit and open-circuit terminals conditions are utilized, hence this parameter representation is known as hybrid parameter representation. In this form of representation, the voltage of the input port and the current of the output port are expressed in terms of the current of the input port and the voltage of the output port.

$$\text{Here, } \begin{bmatrix} V_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} h_{11} & h_{12} \\ h_{21} & h_{22} \end{bmatrix} \begin{bmatrix} I_1 \\ V_2 \end{bmatrix} \quad \text{--- (1)}$$

$$\Rightarrow V_1 = h_{11}I_1 + h_{12}V_2 \quad \text{--- (2)}$$

$$\text{and } I_2 = h_{21}I_1 + h_{22}V_2 \quad \text{--- (3)}$$

Assuming short-circuit conditions at the output $V_2 = 0$. This gives from eqn (2) & (3)

$$h_{11} = \frac{V_1}{I_1}$$

$$h_{21} = \frac{I_2}{I_1}$$

$$\text{Here } h_{11} = \frac{V_1}{I_1} \Big|_{V_2=0} \quad \& \quad h_{22} = \frac{I_2}{V_2} \Big|_{I_1=0}$$

These are called impedance and forward current gain. h_{11} is expressed in ohms while h_{21} is a unitless quantity.

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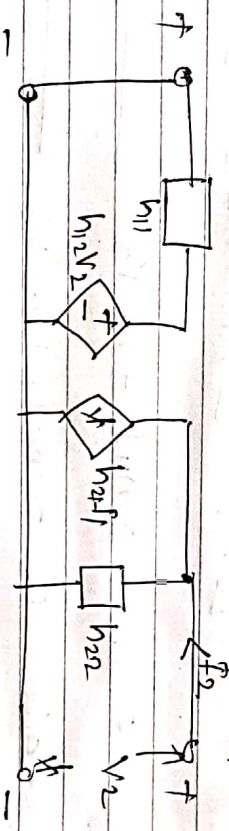
In similar way, from eqn (2) & (13) for input ckt condition open-ckt condition with $I_1 = 0$

$$h_{12} = \left. \frac{V_1}{V_2} \right|_{I_1=0}$$

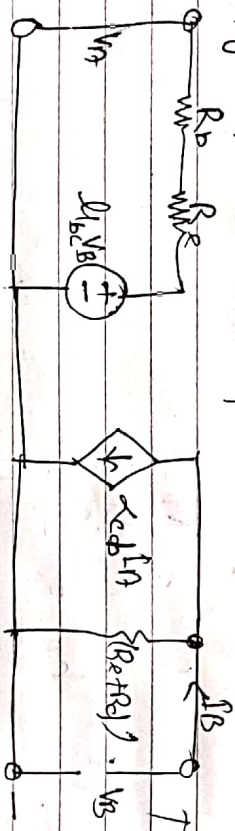
$$\& h_{22} = \left. \frac{I_2}{V_2} \right|_{I_1=0}$$

These are called voltage gain and output admittance h_2 has got no unit while h_{22} has unit of mho.

The equivalent ckt of hybrid param



Q:- CE transistor equivalent ckt has been shown in fig. find the h -parameters



Apply KVL at the input part.

$$V_A = (R_b + R_e)I_A + \beta I_B \quad \text{--- (1)}$$

14

Apply KCL at output part loop

$$I_B = \alpha_{cb} I_A + \frac{V_B}{R_c + R_d} \quad \text{--- (2)}$$

Write (1) & (2) in matrix form.

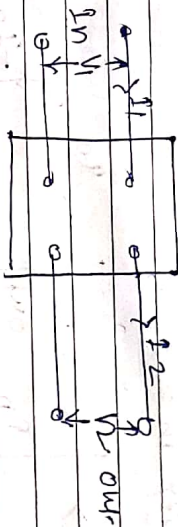
$$\begin{bmatrix} V_A \\ I_B \end{bmatrix} = \begin{bmatrix} R_b + R_e & \beta I_B \\ \alpha_{cb} & \frac{1}{R_c + R_d} \end{bmatrix} \begin{bmatrix} I_A \\ V_B \end{bmatrix}$$

i.e $h_{11} = R_b + R_e$, $h_{12} = \beta I_B$

$$h_{21} = \alpha_{cb}, \quad h_{22} = \frac{1}{R_c + R_d}$$

ABCD parameters

ABCD parameters are widely used in analysis of power transmission engineering where they are termed as circuit parameters. ABCD parameters are also called as Transmission Parameters. It is conventional to designate the input port as sending end and output port as receiving end while representing ABCD parameters. Moreover, the output current direction, assumed in fig is taken reverse.



Hence, the ABCD parameter equation are

$$\text{given as } \begin{bmatrix} V_1 \\ I_1 \end{bmatrix} = \begin{bmatrix} A & B \\ C & D \end{bmatrix} \begin{bmatrix} V_2 \\ -I_2 \end{bmatrix} \quad \text{--- (1)}$$

such that

$$V_1 = AV_2 + B(-I_2) \quad \text{--- (2)}$$

$$I_1 = CV_2 + D(-I_2) \quad \text{--- (3)}$$

Assuming receiving end to be open ckt, $I_2 = 0$

$$\therefore A = \frac{V_1}{V_2} \quad B = \frac{I_1}{V_2}$$

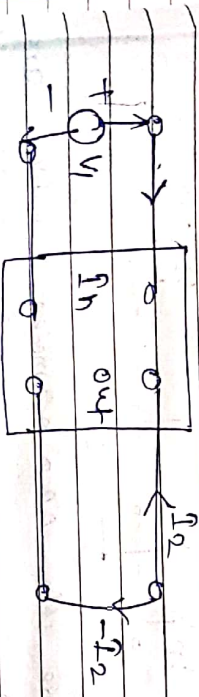
Being a ratio of voltages 'A' is called reverse voltage ratio. It has no unit. 'C' is known as transfer admittance and has the unit mho.

Let us assume that the receiving end short-circuit from eqn (I) & (II) with $V_2 = 0$

$$B = \frac{V_1}{-I_2} \quad \text{or} \quad D = \frac{I_1}{-I_2}$$

D being a ratio of two currents, it is called reverse current ratio & its unitless and is expressed in ohm and is termed as transfer impedance.

Reciprocity in Z-parameter



Let the Z-parameter based two port network be represented as shown in fig. Here from the governing Z-parameter network.

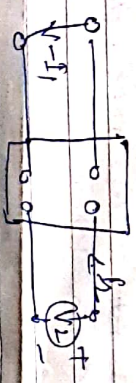
$$V_1 = Z_{11}I_1 + Z_{12}I_2 \quad \text{--- (I)}$$

$$0 = Z_{21}I_1 + Z_{22}I_2 \quad \text{--- (II)}$$

From eqn (I) & (II)

$$I_2 = \frac{V_1 Z_{21}}{Z_{11}Z_{22} - Z_{12}Z_{21}} \quad \text{--- (III)}$$

Let us now interchange the input and output voltage & currents



Here from the governing equation of Z parameter network.

$$V_1 = -Z_{11}I_1 + Z_{12}I_2 \quad \text{--- (I)}$$

$$V_2 = -Z_{21}I_1 + Z_{22}I_2 \quad \text{--- (II)}$$

From eqn (I) & (II)

$$I_1 = \frac{V_2 Z_{12}}{Z_{11}Z_{22} - Z_{12}Z_{21}} \quad \text{--- (III)}$$

Assuming $V_1 = V_2$ comparison between (I) & (III)

$$Z_{12} = Z_{21}$$

∴ This represents the condition of reciprocity.

Symmetry in Z-parameter

Applying voltage V at input port and keeping the output port open.

$$V = Z_{11}I_1 \quad \text{i.e.} \quad Z_{11} = \frac{V}{I_1} \Big|_{I_2=0}$$

Again applying voltage V at output port and keeping open circuit at input port

$$V = Z_{22}I_2 \quad \text{i.e.} \quad Z_{22} = \frac{V}{I_2} \Big|_{I_1=0}$$

condition of symmetry is observed when $V/I_1 = V/I_2$ leading to the condition

$$Z_{11} = Z_{22}$$

∴ This is condition for symmetry.

Condition of Reciprocity and symmetry

Parameter	Condition for reciprocity	Condition for symmetry
Z	$Z_{12} = Z_{21}$	$Z_{11} = Z_{22}$
Y	$Y_{12} = Y_{21}$	$Y_{11} = Y_{22}$
h	$h_{12} = -h_{21}$	$\Delta h = 1$
ABCD	$AD - BC = 1$	$A = D$

Inter-Relationships between parameters of 2-port network:-

① Z-Parameters in terms of Y-parameters.

$$[Z] = [Y]^{-1}$$

$$\Rightarrow \begin{bmatrix} Z_{11} & Z_{12} \\ Z_{21} & Z_{22} \end{bmatrix} = \begin{bmatrix} Y_{11} & Y_{12} \\ Y_{21} & Y_{22} \end{bmatrix}^{-1}$$

$$\text{Thus } Z_{11} = \frac{Y_{22}}{\Delta Y} \quad , \quad Z_{12} = -\frac{Y_{12}}{\Delta Y}$$

$$Z_{21} = -\frac{Y_{21}}{\Delta Y} \quad \& \quad Z_{22} = \frac{Y_{11}}{\Delta Y}$$

$$\text{Here } \Delta Y = \begin{vmatrix} Y_{11} & Y_{12} \\ Y_{21} & Y_{22} \end{vmatrix} = Y_{11}Y_{22} - Y_{12}Y_{21}$$

②

Z-parameters in terms of ABCD parameters from the governing equation

$$I_1 = CI_2 - DI_2$$

$$\Rightarrow V_1 = \frac{1}{C}I_1 + \frac{D}{C}I_2 \quad \text{--- (1)}$$

similarly

$$V_1 = AV_2 - BI_2$$

$$\Rightarrow V_1 = \left[\frac{1}{C}I_1 + \frac{D}{C}I_2 \right] A - BI_2$$

$$\Rightarrow V_1 = \frac{A}{C}I_1 + \frac{AD - BC}{C}I_2 \quad \text{--- (1')}$$

comparing eqn (1) & (1') with standard of Z-parameters

$$\therefore Z_{11} = \frac{A}{C} \quad , \quad Z_{12} = \frac{AD - BC}{C}$$

$$Z_{21} = \frac{1}{C} \quad , \quad Z_{22} = \frac{D}{C}$$

③ Z-parameters in terms of hybrid parameters.

$$\text{From h-parameters equations}$$

$$V_1 = h_{11}I_1 + h_{12}V_2 \quad \text{--- (1)}$$

From eqn (1)

$$V_2 = -\frac{h_{21}}{h_{22}}I_1 + \frac{1}{h_{22}}V_1 \quad \text{--- (2)}$$

From eqn (2)

$$V_1 = h_{11}I_1 + h_{12}V_2$$

$$= h_{11}I_1 + h_{12} \left[-\frac{h_{21}}{h_{22}}I_1 + \frac{1}{h_{22}}V_1 \right] \quad \text{--- (3)}$$

Ans Comparing eqn (8) & (10) with standard form of Z -parameter.

$$Z_{11} = \frac{A}{h_{22}}, \quad Z_{12} = \frac{h_{12}}{h_{22}}$$

$$Z_{21} = -\frac{h_{21}}{h_{22}}, \quad Z_{22} = \frac{1}{h_{22}}$$

$$\text{Here } \Delta h = h_{11}h_{22} - h_{21}h_{12}$$

(4) Y -parameters in terms of Z -parameter

$$\text{We know that } [I] = [Y][V] = [Y][Z][I]$$

$$[Y] = [Z]^{-1} \quad \therefore [V] = [Z][I]$$

$$\Rightarrow \begin{bmatrix} Y_{11} & Y_{12} \\ Y_{21} & Y_{22} \end{bmatrix} = \begin{bmatrix} Z_{11} & Z_{12} \\ Z_{21} & Z_{22} \end{bmatrix}^{-1}$$

$$\text{Thus } Y_{11} = \frac{Z_{22}}{\Delta Z}, \quad Y_{12} = -\frac{Z_{12}}{\Delta Z}$$

$$Y_{21} = -\frac{Z_{21}}{\Delta Z}, \quad Y_{22} = \frac{Z_{11}}{\Delta Z}$$

$$\text{Here, } \Delta Z = Z_{11}Z_{22} - Z_{12}Z_{21}$$

(5) Y -parameters in terms of Transmission parameters

From ABCD parameter

$$V_1 = AV_2 - B I_2$$

$$\Rightarrow I_2 = -\frac{1}{B}V_1 + \frac{A}{B}V_2 \quad \text{--- (1)}$$

$$I_1 = C V_2 - D I_2$$

$$= \left(\frac{D}{B}\right)V_1 - \frac{AD-B^2}{B}V_2 \quad \text{--- (2)}$$

Comparing equations (1) & (2) with standard form of Y -parameter

$$Y_{11} = \frac{D}{B}, \quad Y_{12} = -\frac{AD-B^2}{B}$$

$$Y_{21} = -\frac{C}{B}, \quad Y_{22} = \frac{A}{B}$$

(6) Y -parameters in terms of h -parameter.

From h -parameter equation.

$$V_1 = h_{11}I_1 + h_{12}V_2$$

$$\Rightarrow I_1 = \left(\frac{1}{h_{11}}\right)V_1 + \left(-\frac{h_{12}}{h_{11}}\right)V_2 \quad \text{--- (1)}$$

$$I_2 = h_{21}I_1 + h_{22}V_2$$

$$\Rightarrow I_2 = \frac{h_{21}}{h_{11}}V_1 + \frac{h_{11}h_{22} - h_{12}h_{21}}{h_{11}}V_2 \quad \text{--- (2)}$$

Comparing these two equations with the standard form of Y -parameter network

$$Y_{11} = \frac{1}{h_{11}}, \quad Y_{12} = -\frac{h_{12}}{h_{11}}$$

$$Y_{21} = \frac{h_{21}}{h_{11}}, \quad Y_{22} = \frac{h_{11}h_{22} - h_{12}h_{21}}{h_{11}}$$

Q1: ABCD parameter in terms of Z-parameter.

The Z-parameter equations are given by
 $V_2 = Z_{21}I_1 + Z_{22}I_2$

$$\therefore I_1 = \left(\frac{1}{Z_{21}}\right)V_2 + \left(\frac{Z_{22}}{Z_{21}}\right)I_2 \quad \text{--- (1)}$$

$$\& V_1 = Z_{11}I_1 + Z_{12}I_2$$

$$\therefore V_1 = \left[Z_{11} \left(\frac{1}{Z_{21}} \right) V_2 + \left(\frac{Z_{22}}{Z_{21}} \right) I_2 \right] + Z_{12}I_2$$

$$\Rightarrow V_1 = \left(\frac{Z_{11}}{Z_{21}} \right) V_2 + (-I_2) \left[\frac{Z_{22}Z_{11} - Z_{12}Z_{21}}{Z_{21}} \right] \quad \text{--- (11)}$$

comparing eqn (1) with ABCD parameter equation.

$$A = \frac{Z_{11}}{Z_{21}}, \quad B = - \frac{Z_{22}Z_{11} - Z_{12}Z_{21}}{Z_{21}}$$

$$C = \frac{1}{Z_{21}}, \quad D = \frac{Z_{22}}{Z_{21}}$$

(8) ABCD parameters in terms of Y-parameter
 From Y-parameter

$$I_2 = Y_{21}V_1 + Y_{22}V_2$$

$$\Rightarrow V_1 = \left(-\frac{Y_{22}}{Y_{21}} \right) V_2 + \left(-\frac{1}{Y_{21}} \right) (-I_2)$$

$$\& \text{and } I_1 = Y_{11}V_1 + Y_{12}V_2 \quad \text{--- (12)}$$

Put the value of V_1 in eqn (12)

$$I_1 = Y_{11} \left[\left(-\frac{Y_{22}}{Y_{21}} \right) V_2 + \left(-\frac{1}{Y_{21}} \right) (-I_2) \right] + Y_{12}V_2$$

$$= \left[\frac{Y_{11}Y_{22} - Y_{12}Y_{21}}{Y_{21}} \right] V_2 - \left(-\frac{Y_{11}}{Y_{21}} \right) I_2 \quad \text{--- (11)}$$

comparing eqn (1) & (11) with standard form of ABCD parameters.

$$A = \left(-\frac{Y_{22}}{Y_{21}} \right), \quad B = -\frac{1}{Y_{21}}$$

$$C = \left[\frac{Y_{11}Y_{22} - Y_{12}Y_{21}}{Y_{21}} \right], \quad D = \left(-\frac{Y_{11}}{Y_{21}} \right)$$

(9) ABCD parameters in terms of h-parameter
 From h-parameter equation.

$$I_2 = h_{21}I_1 + h_{22}V_2$$

$$\Rightarrow I_1 = \left(-\frac{h_{22}}{h_{21}} \right) V_2 + \left(-\frac{1}{h_{21}} \right) (-I_2) \quad \text{--- (13)}$$

$$\& V_1 = h_{11}I_1 + h_{12}V_2 \quad \text{--- (14)}$$

Put the value of I_1 in eqn (14)

$$I_1 = \left(-\frac{h_{22}}{h_{21}} \right) V_2 + \left(-\frac{1}{h_{21}} \right) (-I_2) \quad \text{--- (11)}$$

Compare eqn (1) & (11) with standard form of ABCD parameters.

$$A = \left(-\frac{h_{22}}{h_{21}} \right), \quad B = -\left(\frac{h_{11}}{h_{21}} \right)$$

$$C = \left(-\frac{h_{22}}{h_{21}} \right), \quad D = -\left(\frac{1}{h_{21}} \right)$$

10. h-parameters in terms of Z-parameters
From Z-parameters eqn.

$$V_2 = Z_{21}I_1 + Z_{22}I_2$$

$$\Rightarrow I_2 = \left(-\frac{Z_{21}}{Z_{22}}\right)I_1 + \left(\frac{1}{Z_{22}}\right)V_2 \quad \text{--- (1)}$$

$$\text{and } V_1 = Z_{11}I_1 + Z_{12}I_2 \quad \text{--- (11)}$$

put the value of I_2 in eqn (11)

$$V_1 = Z_{11}I_1 + Z_{12} \left[\left(-\frac{Z_{21}}{Z_{22}}\right)I_1 + \left(\frac{1}{Z_{22}}\right)V_2 \right]$$

Comparing eqn (I) & (II) with standard form of h-parameters. (14)

$$h_{11} = \frac{Z_{11}Z_{22} - Z_{21}Z_{12}}{Z_{22}}$$

$$h_{12} = \frac{Z_{12}}{Z_{22}}$$

$$h_{21} = -\frac{Z_{21}}{Z_{22}}$$

$$\text{and } h_{22} = \frac{1}{Z_{22}}$$

11. h-parameters in terms of the Y-parameters
From Y-parameters eqn

$$I_1 = Y_{11}V_1 + Y_{12}V_2 \quad \text{--- (1)}$$

$$I_2 = Y_{21}V_1 + Y_{22}V_2 \quad \text{--- (11)}$$

$$\rightarrow V_1 = \left(\frac{1}{Y_{11}}\right)I_1 + \left(-\frac{Y_{12}}{Y_{11}}\right)V_2 \quad \text{--- (11)}$$

put V_1 in eqn (11)

$$I_2 = Y_{21}V_1 + Y_{22}V_2$$

$$= \left(\frac{Y_{21}}{Y_{11}}\right)I_1 + \left(\frac{Y_{11}Y_{22} - Y_{12}Y_{21}}{Y_{11}}\right)V_2 \quad \text{--- (12)}$$

comparing eqn (11) & (12) with standard form of h-parameters.

$$h_{11} = \frac{1}{Y_{11}}$$

$$h_{12} = -\frac{Y_{12}}{Y_{11}}$$

$$h_{21} = \frac{Y_{21}}{Y_{11}}$$

$$h_{22} = \frac{Y_{11}Y_{22} - Y_{12}Y_{21}}{Y_{11}}$$

12) h-parameters in terms of ABCD-parameters.

From ABCD-parameters

$$I_1 = CV_2 - DI_2$$

$$\Rightarrow I_2 = \left(-\frac{1}{D}\right)I_1 + \left(\frac{C}{D}\right)V_2 \quad \text{--- (I)}$$

$$V_1 = AV_2 - BI_2 \quad \text{--- (II)}$$

put value of I_2 in eqn (II)

$$V_1 = AV_2 - B\left[-\frac{1}{D}I_1 + \left(\frac{C}{D}\right)V_2\right]$$

$$= \left(\frac{B}{D}\right)I_1 + \left(\frac{AD - BC}{D}\right)V_2 \quad \text{--- (III)}$$

Comparing (I) & (III) with standard form of h-parameters.

$$h_{11} = \left(\frac{B}{D}\right)$$

$$h_{12} = \left(\frac{AD - BC}{D}\right)$$

$$h_{21} = \left(-\frac{1}{D}\right)$$

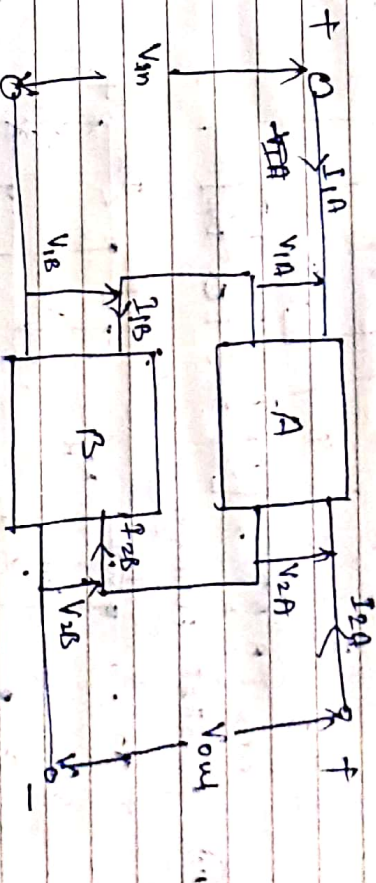
$$h_{22} = \left(\frac{C}{D}\right)$$

Different types of interconnection of two port networks

1) series connection:-

Let network A & B be the two port networks connected in series

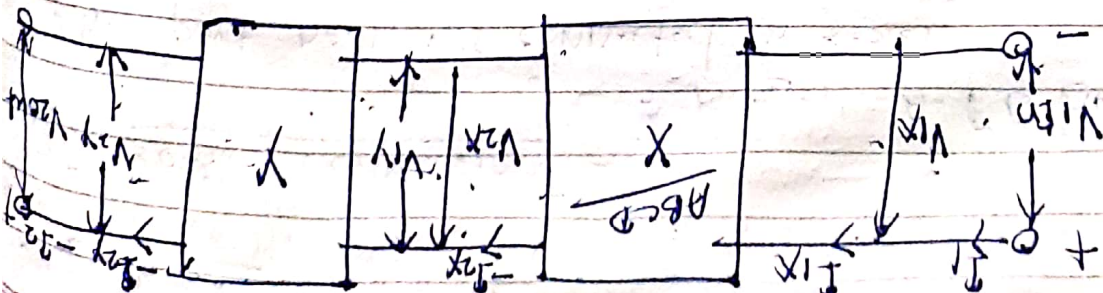
All input and output currents and voltages shown in fig.



In this form connection.

$$\begin{bmatrix} V_1 \\ V_2 \end{bmatrix} = \begin{bmatrix} Z_{11A} + Z_{11B} & Z_{12A} + Z_{12B} \\ Z_{21A} + Z_{21B} & Z_{22A} + Z_{22B} \end{bmatrix} \begin{bmatrix} I_1 \\ I_2 \end{bmatrix}$$

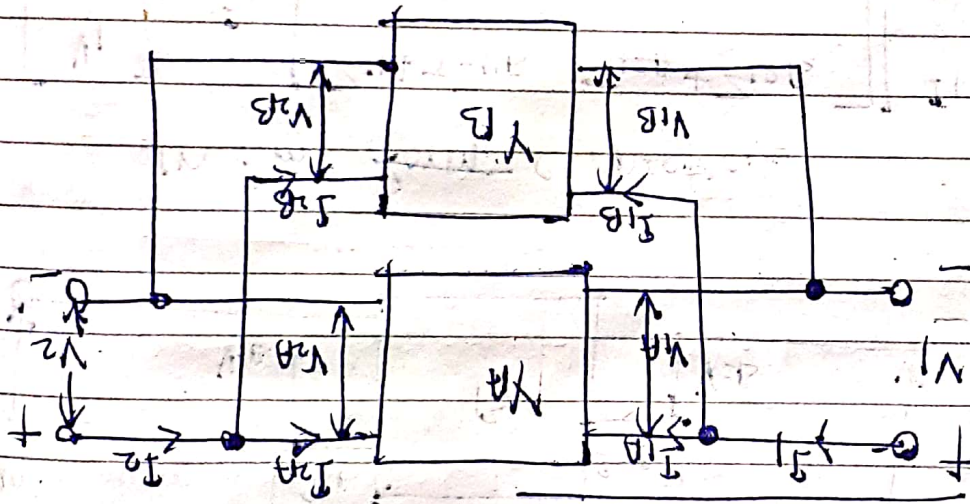
II Cascode connection:



III

Here $\begin{bmatrix} A & B \\ C & D \end{bmatrix} = \begin{bmatrix} A_x & B_x \\ C_x & D_x \end{bmatrix} \begin{bmatrix} A_y & B_y \\ C_y & D_y \end{bmatrix}$

parallel connection



II

$\begin{bmatrix} I_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} Y_{11A} + Y_{11B} & Y_{12A} + Y_{12B} \\ Y_{21A} + Y_{21B} & Y_{22A} + Y_{22B} \end{bmatrix} \begin{bmatrix} V_1 \\ V_2 \end{bmatrix}$